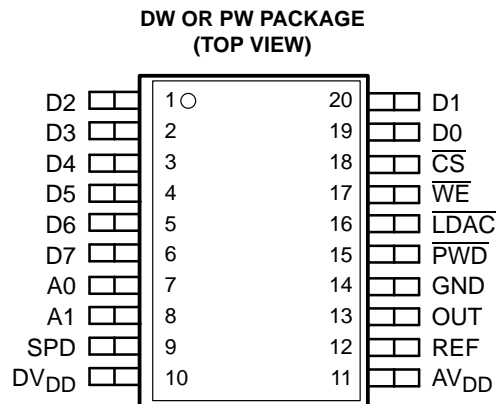


TLV5613

2.7 V TO 5.5 V 12-BIT PARALLEL DIGITAL-TO-ANALOG CONVERTER WITH POWER DOWN

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- 12-Bit Voltage Output DAC
- Single Supply 2.7-V to 5.5-V Operation
- Separate Analog and Digital Supplies
- ± 0.4 LSB Differential Nonlinearity (DNL),
 ± 1.5 LSB Integral Nonlinearity (INL)
- Programmable Settling Time vs Power Consumption
 - 1 μ s/8mW in Fast Mode
 - 3.5 μ s/1.2 mW in Slow Mode
- 8-Bit μ Controller Compatible Interface (8+4 Bit)
- Power-Down Mode (50 nW)
- Rail-to-Rail Output Buffer
- Synchronous or Asynchronous Update
- Monotonic Over Temperature



applications

- Digital Servo Control Loops
- Battery Powered Test Instruments
- Digital Offset and Gain Adjustment
- Industrial Process Control
- Speech Synthesis
- Machine and Motion Control Devices
- Mass Storage Devices

description

The TLV5613 is a 12-bit voltage output digital-to-analog converter (DAC) with a 8-bit microcontroller compatible parallel interface. The 8 LSBs, the 4 MSBs and 5 control bits are written using three different addresses. Developed for a wide range of supply voltages, the TLV5633 can be operated from 2.7 V to 5.5 V.

The resistor string output voltage is buffered by a x2 gain rail-to-rail output buffer. The buffer features a Class A (slow mode: AB) output stage to improve stability and reduce settling time. The programmable settling time of the DAC allows the designer to optimize speed versus power dissipation. With its on chip programmable precision voltage reference, the TLV5613 simplifies overall system design. The settling time and the reference voltage can be chosen by the control bits within the 16 bit data word.

Implemented with a CMOS process, the device is designed for single supply operation from 2.7 V to 5.5 V. It is available in 20 pin SOIC in standard commercial and industrial temperature ranges.

AVAILABLE OPTIONS

PACKAGE		
T _A	SMALL OUTLINE (DW)	TSSOP (PW)
0°C to 70°C	TLV5613CDW	TLV5613CPW
-40°C to 85°C	TLV5613IDW	TLV5613IPW

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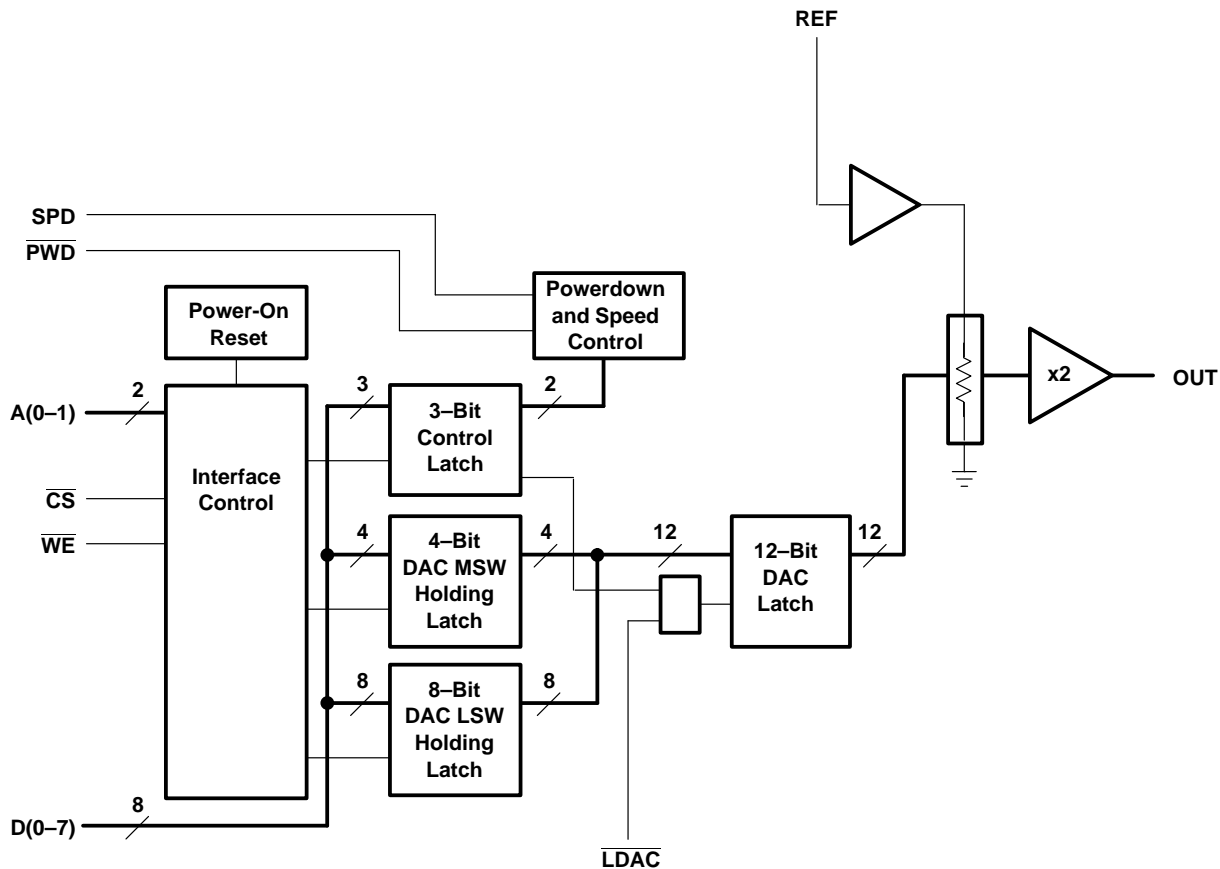
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functional block diagram



Terminal Functions

TERMINAL NAME	NO.	I/O	DESCRIPTION
AV _{DD}	11		Analog positive power supply
A0	7	I	Address input
A1	8	I	Address input
$\overline{\text{CS}}$	18	I	Chip select. Digital input active low, used to enable/disable inputs
DV _{DD}	10		Digital positive power supply
D0 (LSB) – D7 (MSB)	1–6, 19, 20	I	Data input
$\overline{\text{LDAC}}$	16	I	Load DAC. Digital input active low, used to load DAC output
OUT	13	O	DAC analog voltage output
$\overline{\text{PWD}}$	15	I	Power down. Digital input active low
REF	12	I	Analog reference voltage input
SPD	9	I	Speed select. Digital input
GND	14		Ground
$\overline{\text{WE}}$	17	I	Write enable. Digital input active low, used to latch data

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absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Supply voltage (DV_{DD} , AV_{DD} to GND)	7 V
Supply voltage difference AV_{DD} to DV_{DD}	– 2.8 V to 2.8 V
Reference input voltage range	– 0.3 V to $AV_{DD} + 0.3$ V
Digital input voltage range to GND	– 0.3 V to $DV_{DD} + 0.3$ V
Operating free-air temperature range, T_A : TLV5613C	0°C to 70°C
TLV5613I	–40°C to 85°C
Storage temperature range, T_{stg}	–65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

† Stresses beyond those listed under “absolute maximum ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under “recommended operating conditions” is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, V_{DD}	5-V Supply	4.5	5	5.5	V
	3-V Supply	2.7	3	3.3	
Supply voltage difference $\Delta V_{DD} = AV_{DD} - DV_{DD}$		–2.8	0	2.8	V
Power on reset, POR		0.55		2	V
High-level digital input voltage, V_{IH}	$DV_{DD} = 2.7$ V to 5.5V	2			V
Low-level digital input voltage, V_{IL}	$DV_{DD} = 2.7$ V to 5.5V			0.8	V
Reference voltage, V_{ref} to REFIN terminal	5-V Supply (see Note 1)	GND	2.048	$AV_{DD} - 1.5$	V
	3-V Supply (see Note 1)	GND	1.024	$AV_{DD} - 1.5$	
Load resistance, R_L		2			k Ω
Load capacitance, C_L				100	pF
Operating free-air temperature, T_A	TLV5613C	0		70	°C
	TLV5613I	–40		85	°C

NOTE 1: Due to the x2 output buffer, a reference input voltage $\geq V_{DD}/2$ causes clipping of the transfer function.

PRODUCT PREVIEW



2.7 V TO 5.5 V 12-BIT PARALLEL DIGITAL-TO-ANALOG CONVERTER WITH POWER DOWN

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electrical characteristics over recommended operating free-air temperature range, supply voltages, and reference voltages (unless otherwise noted)

power supply

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
I _{DD}	Power supply current	No load, All inputs = GND or DV _{DD} , DAC latch = 0x800	V _{DD} = 5 V	Fast	1.6	3	mA
				Slow	0.5	1.3	mA
			V _{DD} = 3 V	Fast	1.4	2.7	mA
				Slow	0.4	1.1	mA
Power down supply current		AV _{DD} = 5 V		0.01	10	μA	
		AV _{DD} = 3 V		0.01	10		
PSRR	Power supply rejection ratio	Zero scale	See Note 2	-65		dB	
		Full scale	See Note 3	-65			

NOTES: 2. Power supply rejection ratio at zero scale is measured by varying V_{DD} and is given by:

$$PSRR = 20 \log [(E_{ZS}(V_{DDmax}) - E_{ZS}(V_{DDmin})) / V_{DDmax}]$$

3. Power supply rejection ratio at full scale is measured by varying V_{DD} and is given by:

$$PSRR = 20 \log [(E_G(V_{DDmax}) - E_G(V_{DDmin})) / V_{DDmax}]$$

static DAC specifications

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
Resolution		V _{ref} (REFIN) = 2.048V, 1.024V		12			bits
Integral nonlinearity (INL), end point adjusted		V _{ref} (REFIN) = 2.048V, 1.024V, See Note 4			±1.5	±4	LSB
Differential nonlinearity (DNL)		V _{ref} (REFIN) = 2.048V, 1.024V, See Note 5			±0.4	±1	LSB
E _{ZS}	Zero-scale error (offset error at zero scale)	V _{ref} (REFIN) = 2.048V, 1.024V, See Note 6			±3	±20	mV
Zero-scale-error temperature coefficient		V _{ref} (REFIN) = 2.048V, 1.024V, See Note 7			3		ppm/°C
E _G	Gain error	V _{ref} (REFIN) = 2.048V, 1.024V, See Note 8			±0.25	±0.5	% of FS voltage
Gain error temperature coefficient		V _{ref} (REFIN) = 2.048V, 1.024V, See Note 9			1		ppm/°C

NOTES: 4. The relative accuracy or integral nonlinearity (INL) sometimes referred to as linearity error, is the maximum deviation of the output from the line between zero and full scale excluding the effects of zero code and full-scale errors.

5. The differential nonlinearity (DNL) sometimes referred to as differential error, is the difference between the measured and ideal 1 LSB amplitude change of any two adjacent codes. Monotonic means the output voltage changes in the same direction (or remains constant) as a change in the digital input code.

6. Zero-scale error is the deviation from zero voltage output when the digital input code is zero.

7. Zero-scale-error temperature coefficient is given by: $E_{ZS} TC = [E_{ZS}(T_{max}) - E_{ZS}(T_{min})] / V_{ref} \times 10^6 / (T_{max} - T_{min})$.

8. Gain error is the deviation from the ideal output (V_{ref} - 1 LSB) with an output load of 10 kΩ excluding the effects of the zero-error.

9. Gain temperature coefficient is given by: $E_G TC = [E_G(T_{max}) - E_G(T_{min})] / V_{ref} \times 10^6 / (T_{max} - T_{min})$.

output specifications

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
V _O	Output voltage	R _L = 10 kΩ		0		AV _{DD} -0.4	V
Output load regulation accuracy		V _O (OUT) = 4.096 V, 2.048 V, R _L = 2 kΩ,			0.1	0.29	% of FS voltage
I _{OSC} (source)	Output short circuit source current	V _O (OUT) = 0 V, input all 1s	AV _{DD} = 5V	-100			mA
			AV _{DD} = 3V	-25			
I _{OSC} (sink)	Output short circuit sink current	R _L = 100 Ω, input all 1s	AV _{DD} = 5V	-10			mA
			AV _{DD} = 3V	-10			

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electrical characteristics over recommended operating free-air temperature range, supply voltages, and reference voltages (unless otherwise noted)

reference input (REFIN)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V _{ref} Input voltage reference	See Note 10	0		V _{DD} -1.5	V
R _i Input resistance			10		MΩ
C _i Input capacitance			5		pF
Reference input bandwidth	REF = 0.2 V _{pp} + 1.024 V dc	Fast mode	1.6		MHz
		Slow mode	1		MHz
Reference feed through	REF = 1 V _{pp} at 1 kHz + 1.024 V dc (see Note 10)		-60		dB

NOTES: 10. Reference feedthrough is measured at the DAC output with an input code = 0x000.

digital inputs

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
I _{IH} High-level digital input current	V _I = DV _{DD}			1	μA
I _{IL} Low-level digital input current	V _I = 0 V	-1			μA
C _i Input capacitance			8		pF

operating characteristics over recommended operating free-air temperature range, supply voltages, and reference voltages (unless otherwise noted)

analog output dynamic performance

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t _s (FS) Output settling time, full scale	R _L = 10 kΩ, C _L = 100 pF, See Note 11	Fast	1	3	μs
		Slow	3.5	7	
t _s (CC) Output settling time, code to code	R _L = 10 kΩ, C _L = 100 pF, See Note 12	Fast	0.5	1.5	μs
		Slow	1	2	
SR Slew rate	R _L = 10 kΩ, C _L = 100 pF, See Note 13	Fast	8		V/μs
		Slow	1.5		
Glitch energy	DIN = 0 to 1, f _{CLK} = 100 kHz, CS = DV _{DD}		5		nV-S
Signal-to-noise S/N	f _s = 480 kSPS, f _{out} = 1 kHz, R _L = 10 k, C _L = 100 pF		65	78	dB
Signal-to-noise + distortion, S/(N+D)			58	69	
THD Total harmonic distortion			-68	-60	
Spurious free dynamic range			60	72	

NOTES: 11. Settling time is the time for the output signal to remain within ±0.5 LSB of the final measured value for a digital input code change of 0x020 to 0x3FF or 0x3FF to 0x020.

12. Settling time is the time for the output signal to remain within ± 0.5 LSB of the final measured value for a digital input code change of one count. The max time applies to code changes near zero scale or full scale.

13. Slew rate determines the time it takes for a change of the DAC output from 10% to 90% full-scale voltage.

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timing requirements

digital inputs

		MIN	NOM	MAX	UNIT
$t_{su}(D-WE)$	Setup time, data ready before positive \overline{WE} edge	9			ns
$t_{su}(CS-WE)$	Setup time, \overline{CS} low before positive \overline{WE} edge	13			ns
$t_{su}(A-WE)$	Setup time, address bits A0, A1	17			ns
$t_h(D)$	Hold time, data held after positive \overline{WE} edge	0			ns
$t_{su}(WE-LD)$	Setup time, positive \overline{WE} edge before \overline{LDAC} low	0			ns
$t_{wh}(WE)$	Pulse duration, \overline{WE} high	10			ns
$t_w(LD)$	\overline{LDAC} pulse width low	10			μ s

PARAMETER MEASUREMENT INFORMATION

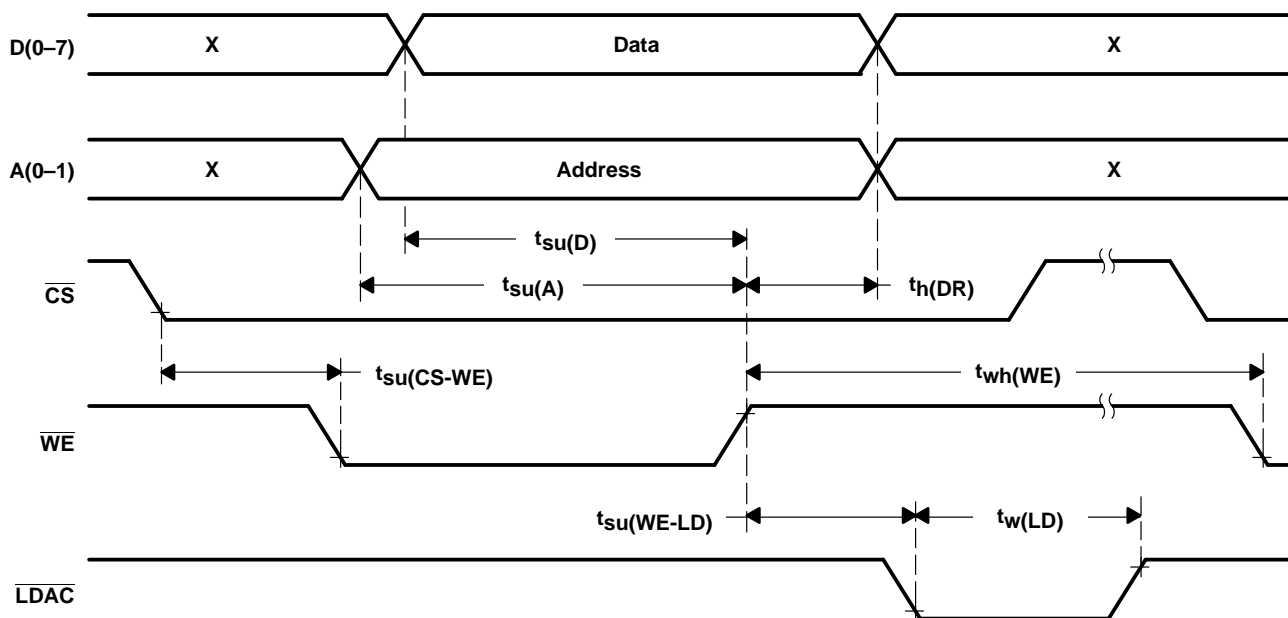


Figure 1. Timing Diagram

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PARAMETER MEASUREMENT INFORMATION

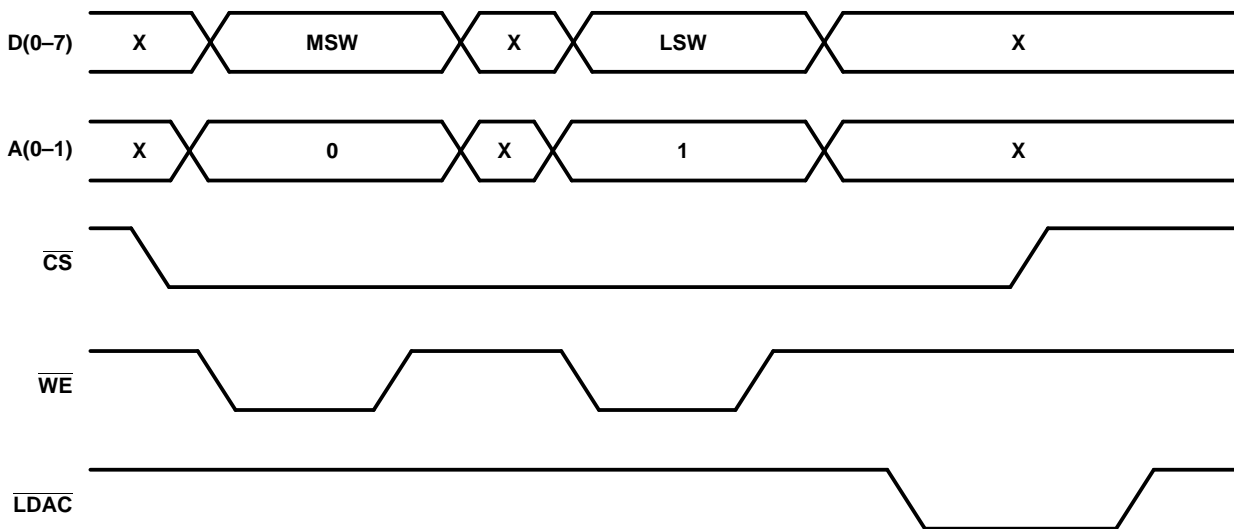


Figure 2. Example of a Complete Write Cycle Using \overline{LDAC} to Update the DAC

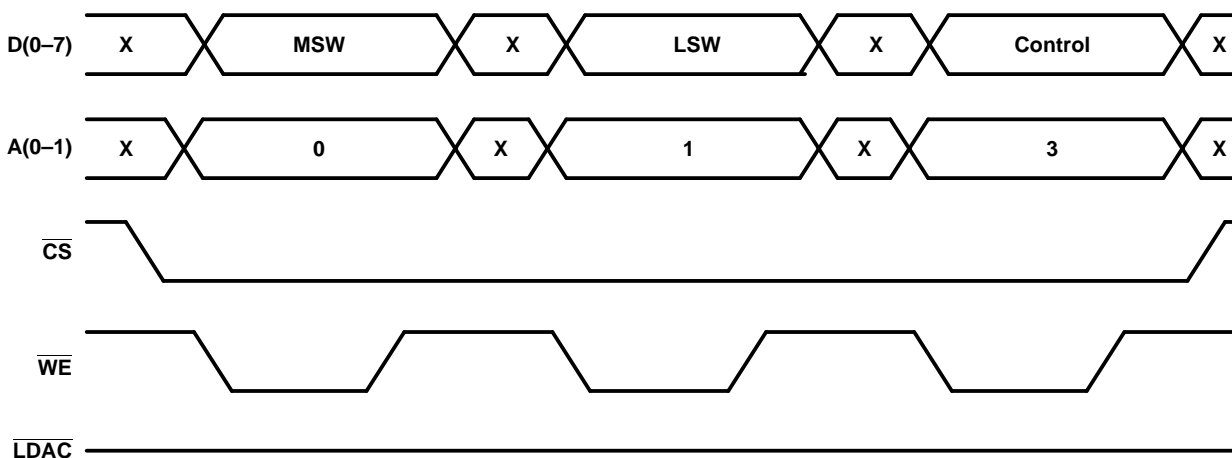


Figure 3. Example of a Complete Write Cycle Using the Control Word to Update the DAC

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TYPICAL CHARACTERISTICS

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**MAXIMUM OUTPUT VOLTAGE
vs
LOAD**

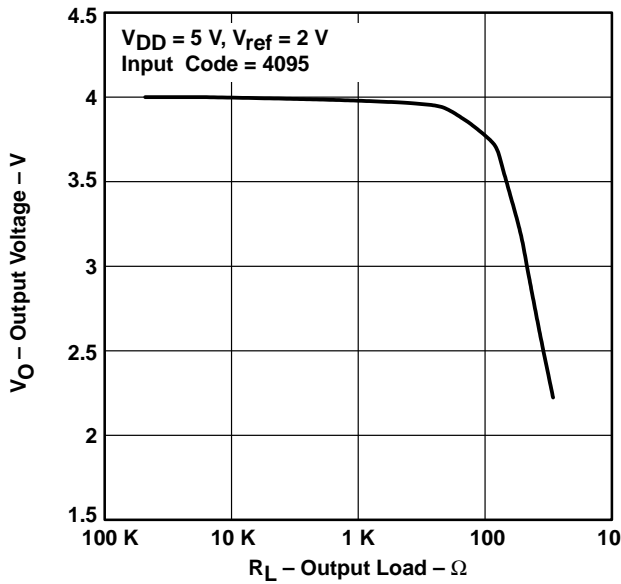


Figure 4

**MAXIMUM OUTPUT VOLTAGE
vs
LOAD**

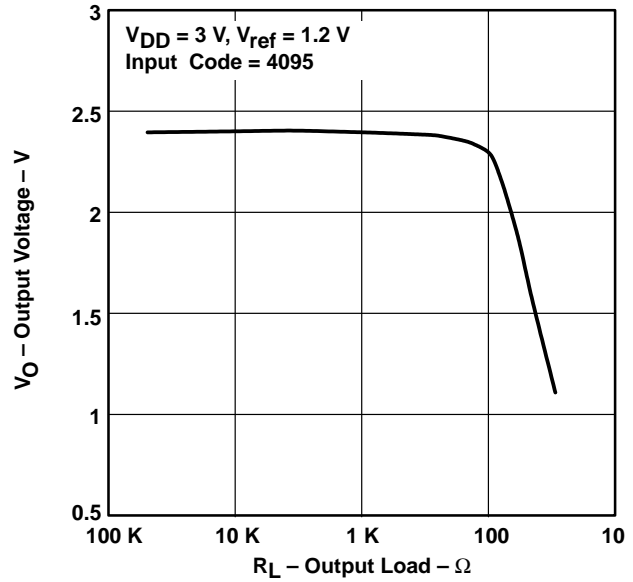


Figure 5

**TOTAL HARMONIC DISTORTION
vs
LOAD**

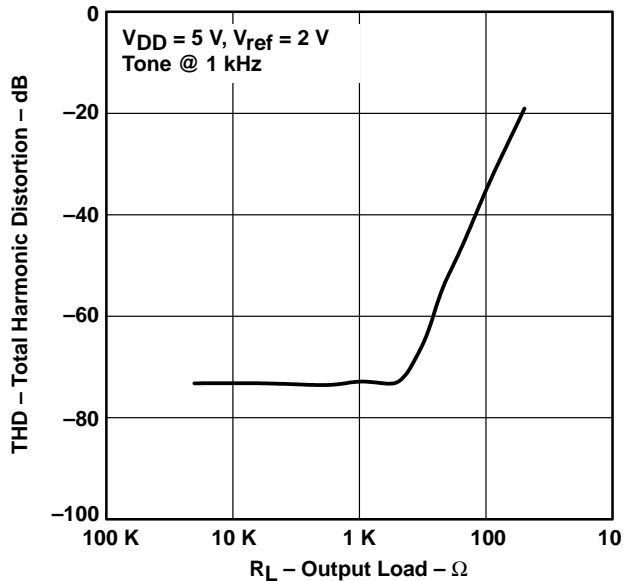


Figure 6

**TOTAL HARMONIC DISTORTION
vs
LOAD**

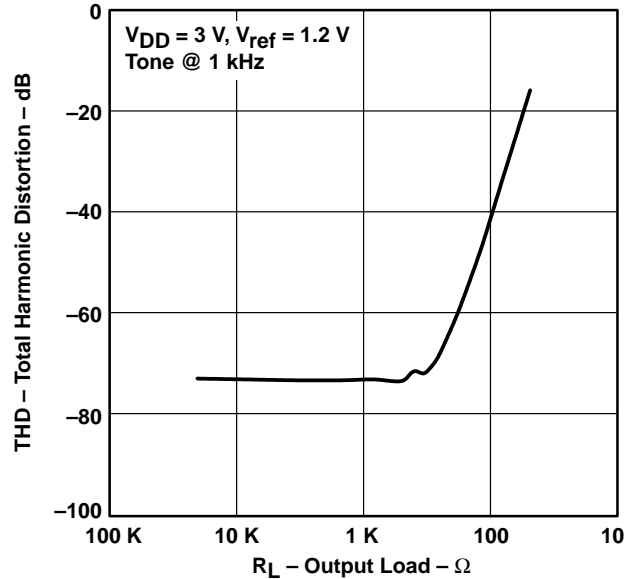


Figure 7



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**2.7 V TO 5.5 V 12-BIT PARALLEL DIGITAL-TO-ANALOG CONVERTER
 WITH POWER DOWN**

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TYPICAL CHARACTERISTICS

**TOTAL HARMONIC DISTORTION
 vs
 FREQUENCY**

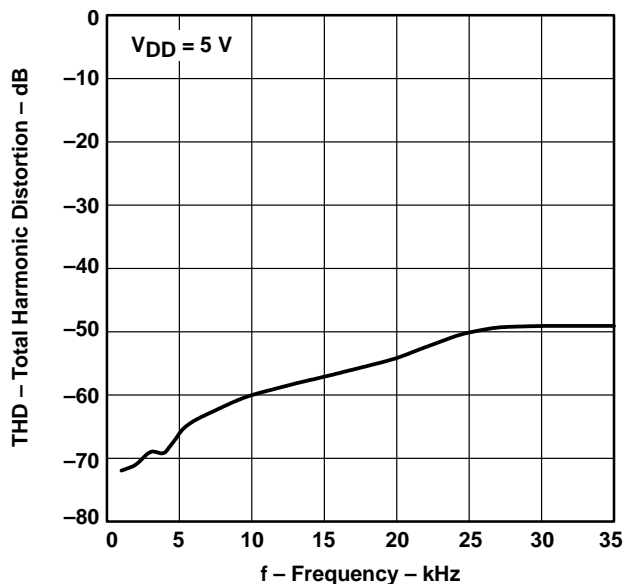


Figure 8

**TOTAL HARMONIC DISTORTION
 vs
 FREQUENCY**

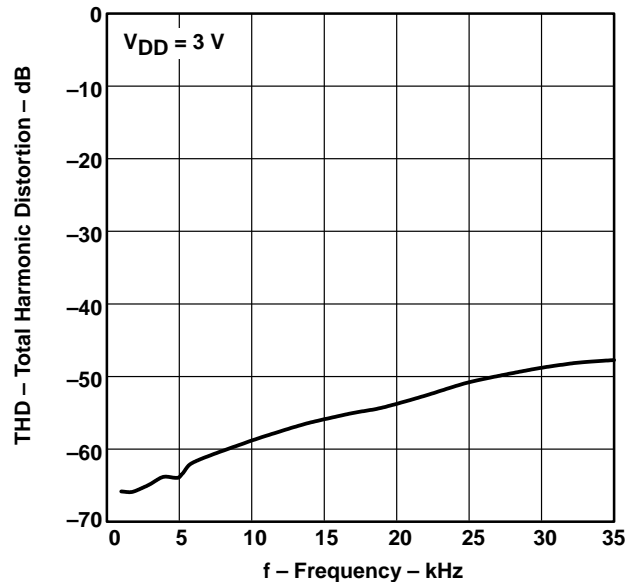


Figure 9

**SIGNAL-TO-NOISE + DISTORTION
 vs
 FREQUENCY**

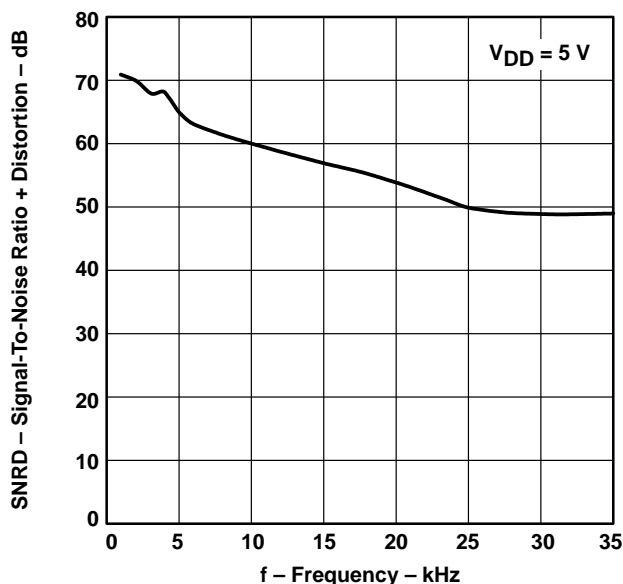


Figure 10

**SIGNAL-TO-NOISE + DISTORTION
 vs
 FREQUENCY**

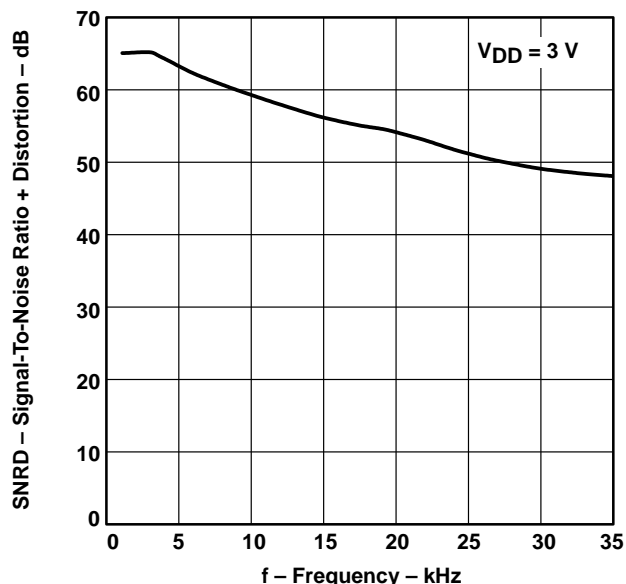


Figure 11

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TYPICAL CHARACTERISTICS

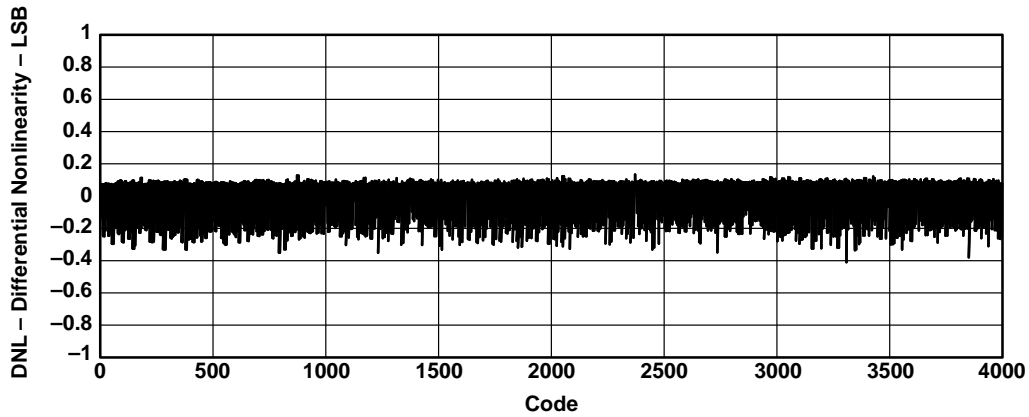


Figure 12. Differential Nonlinearity

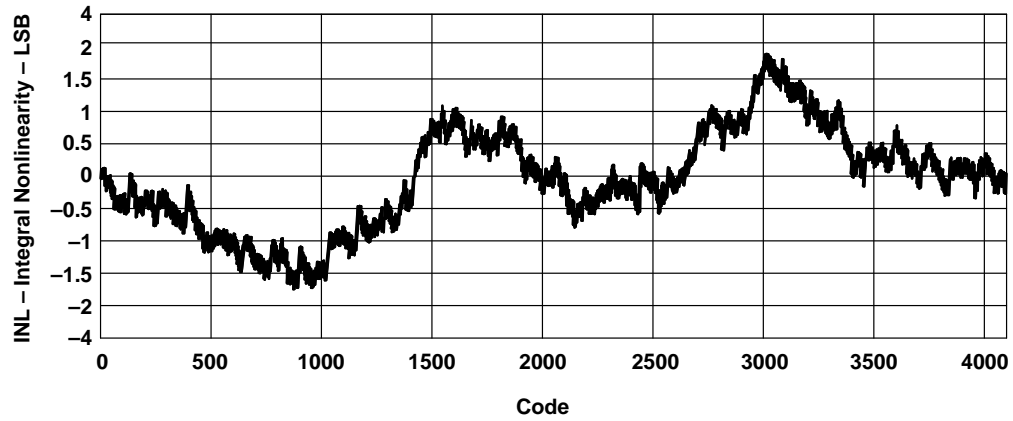


Figure 13. Integral Nonlinearity

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TYPICAL CHARACTERISTICS

**POWER DOWN SUPPLY CURRENT
 vs
 TIME**

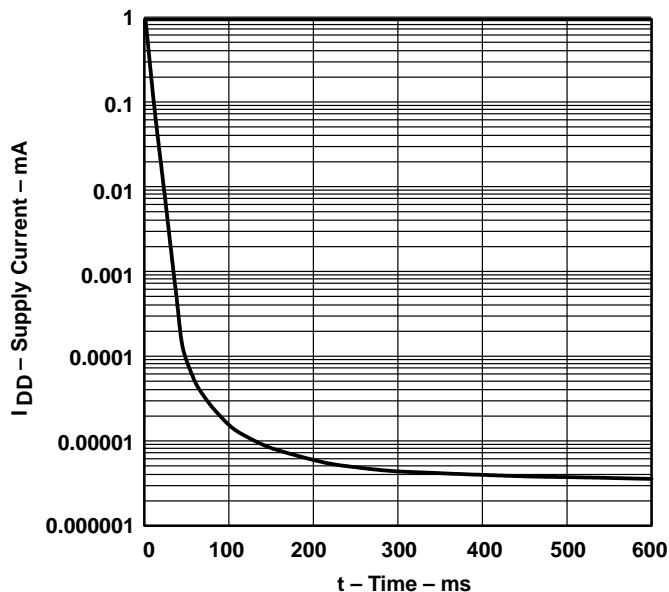


Figure 14

APPLICATION INFORMATION

general function

The TLV5613 is a 12-bit, single supply DAC, based on a resistor string architecture. It consists of a parallel interface, a speed and power down control logic, a programmable internal reference, a resistor string and a rail-to-rail output buffer. The output voltage (full scale determined by reference) is given by:

$$2 \text{ REF } \frac{\text{CODE}}{0x1000} \text{ [V]}$$

Where REF is the reference voltage and CODE is the digital input value, range 0x000 to 0xFFFF. A power on reset initially puts the internal latches to a defined state (all bits zero).

parallel interface

The device latches data on the positive edge of \overline{WE} . It must be enabled with \overline{CS} low. Whether the data is written to one of the DAC holding latches (MSW, LSW) or the control register, depends on the address bits A1 and A0. \overline{LDAC} low updates the DAC with the value in the holding latch. \overline{LDAC} is an asynchronous input and can be held low, if a separate update is not necessary. However, if you want to control the DAC using the load feature, you should wait approximately 20 ns after the positive \overline{WE} edge until driving \overline{LDAC} low. Two more asynchronous inputs, SPD and \overline{PWD} control the settling times and the power down mode:

SPD:	Speed control.	1 → fast mode	0 → slow mode
\overline{PWD} :	Power control.	1 → normal operation	0 → power down

APPLICATION INFORMATION

PIN	BIT	MODE
SPD	SPD	
0	0	Slow
0	1	Fast
1	0	Fast
1	1	Fast

PIN	BIT	POWER
$\overline{\text{PWD}}$	PWD	
0	0	Down
0	1	Down
1	0	Normal
1	1	Down

PIN	BIT	LATCH
$\overline{\text{LDAC}}$	RLDAC	
0	0	Transparent
0	1	Transparent
1	0	Hold
1	1	Transparent

data format

The TLV5613 writes data either to one of the DAC holding latches or to the control register depending on the address bits A1 and A0.

ADDRESS BITS

A1	A0	REGISTER
0	0	DAC LSW holding
0	1	DAC MSW holding
1	0	Reserved
1	1	CONTROL

D7	D6	D5	D4	D3	D2	D1	D0
X	X	X	REF1	REF0	RLDAC	PWD	SPD

X: don't care

SPD: Speed control bit. 1 → fast mode 0 → slow mode
 PWD: Power control bit. 1 → power down 0 → normal operation
 RLDAC: Load DAC latch 1 → latch transparent 0 → DAC latch controlled by $\overline{\text{LDAC}}$ pin

PRODUCT PREVIEW



APPLICATION INFORMATION

layout considerations

To achieve the best performance, it is recommended to have separate power planes for GND, AV_{DD}, and DV_{DD}. Figure 15 shows how to lay out the power planes for the TLV5613. As a general rule, digital and analog signals should be separated as wide as possible. To avoid crosstalk, analog and digital traces must not be routed in parallel. The two positive power planes (AV_{DD} and DV_{DD}) should be connected together at one point with a ferrite bead.

A 100 nF ceramic low series inductance capacitor between DV_{DD} and GND and a 1 μF tantalum capacitor between AV_{DD} and GND as close as possible to the supply pins are recommended for optimal performance.

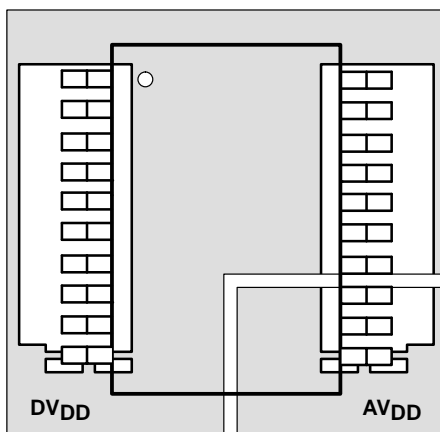


Figure 15. TLV5613 Board Layout

linearity, offset, and gain error using single end supplies

When an amplifier is operated from a single supply, the voltage offset can still be either positive or negative. With a positive offset, the output voltage changes on the first code change. With a negative offset the output voltage may not change with the first code depending on the magnitude of the offset voltage.

The output amplifier attempts to drive the output to a negative voltage. However, because the most negative supply rail is ground, the output cannot drive below ground and clamps the output at 0 V.

The output voltage remains at zero until the input code value produces a sufficient positive output voltage to overcome the negative offset voltage, resulting in the transfer function shown in Figure 11.

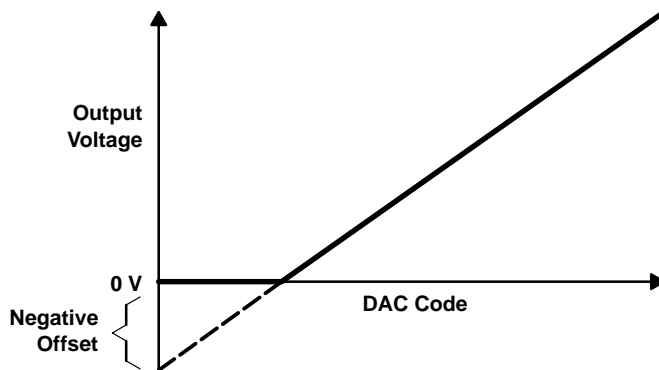


Figure 16. Effect of Negative Offset (Single Supply)

APPLICATION INFORMATION

This offset error, not the linearity error, produces this breakpoint. The transfer function would have followed the dotted line if the output buffer could drive below the ground rail.

For a DAC, linearity is measured between zero input code (all inputs 0) and full scale code (all inputs 1) after offset and full scale are adjusted out or accounted for in some way. However, single supply operation does not allow for adjustment when the offset is negative due to the breakpoint in the transfer function. So the linearity is measured between full scale code and the lowest code that produces a positive output voltage.

definitions of specifications and terminology

integral nonlinearity (INL)

The relative accuracy or integral nonlinearity (INL), sometimes referred to as linearity error, is the maximum deviation of the output from the line between zero and full scale excluding the effects of zero code and full-scale errors.

differential nonlinearity (DNL)

The differential nonlinearity (DNL), sometimes referred to as differential error, is the difference between the measured and ideal 1 LSB amplitude change of any two adjacent codes. Monotonic means the output voltage changes in the same direction (or remains constant) as a change in the digital input code.

zero-scale error (E_{ZS})

Zero-scale error is defined as the deviation of the output from 0 V at a digital input value of 0.

gain error (E_G)

Gain error is the error in slope of the DAC transfer function.

signal-to-noise ratio + distortion (SINAD)

Signal-to-noise ratio + distortion is the ratio of the rms value of the output signal to the rms sum of all other spectral components below the Nyquist frequency, including harmonics but excluding dc. The value for SINAD is expressed in decibels.

spurious free dynamic range (SFDR)

SFDR is the difference between the rms value of the output signal and the rms value of the spurious signal within a specified bandwidth. The value for SFDR is expressed in decibels.

total harmonic distortion (THD)

THD is the ratio of the rms sum of the first six harmonic components to the rms value of the fundamental signal and is expressed in decibels.

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 WITH POWER DOWN**

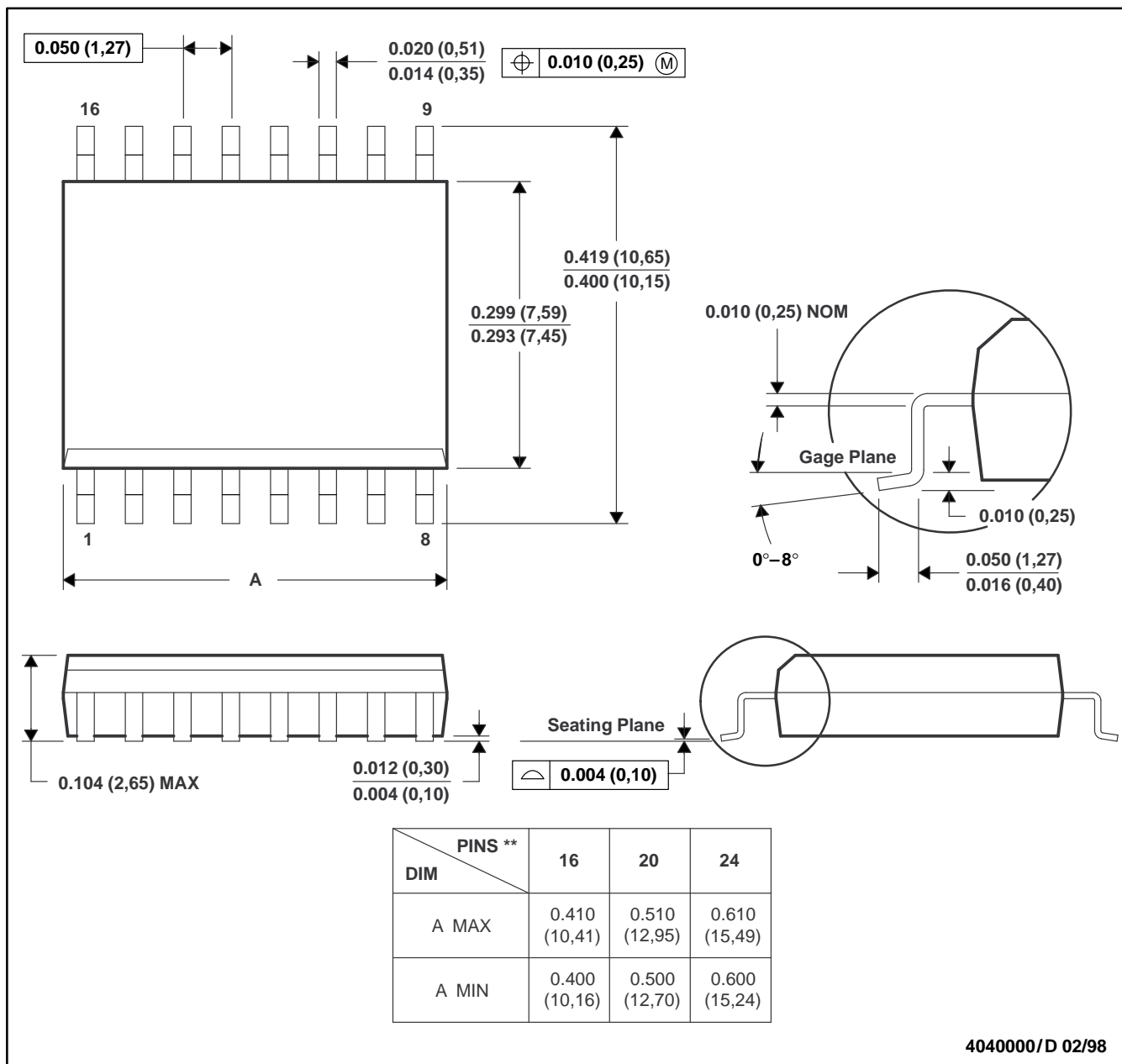
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MECHANICAL DATA

DW (R-PDSO-G)**

PLASTIC SMALL-OUTLINE PACKAGE

16 PIN SHOWN



- NOTES: A. All linear dimensions are in inches (millimeters).
 B. This drawing is subject to change without notice.
 C. Body dimensions do not include mold flash or protrusion not to exceed 0.006 (0,15).
 D. Falls within JEDEC MS-013



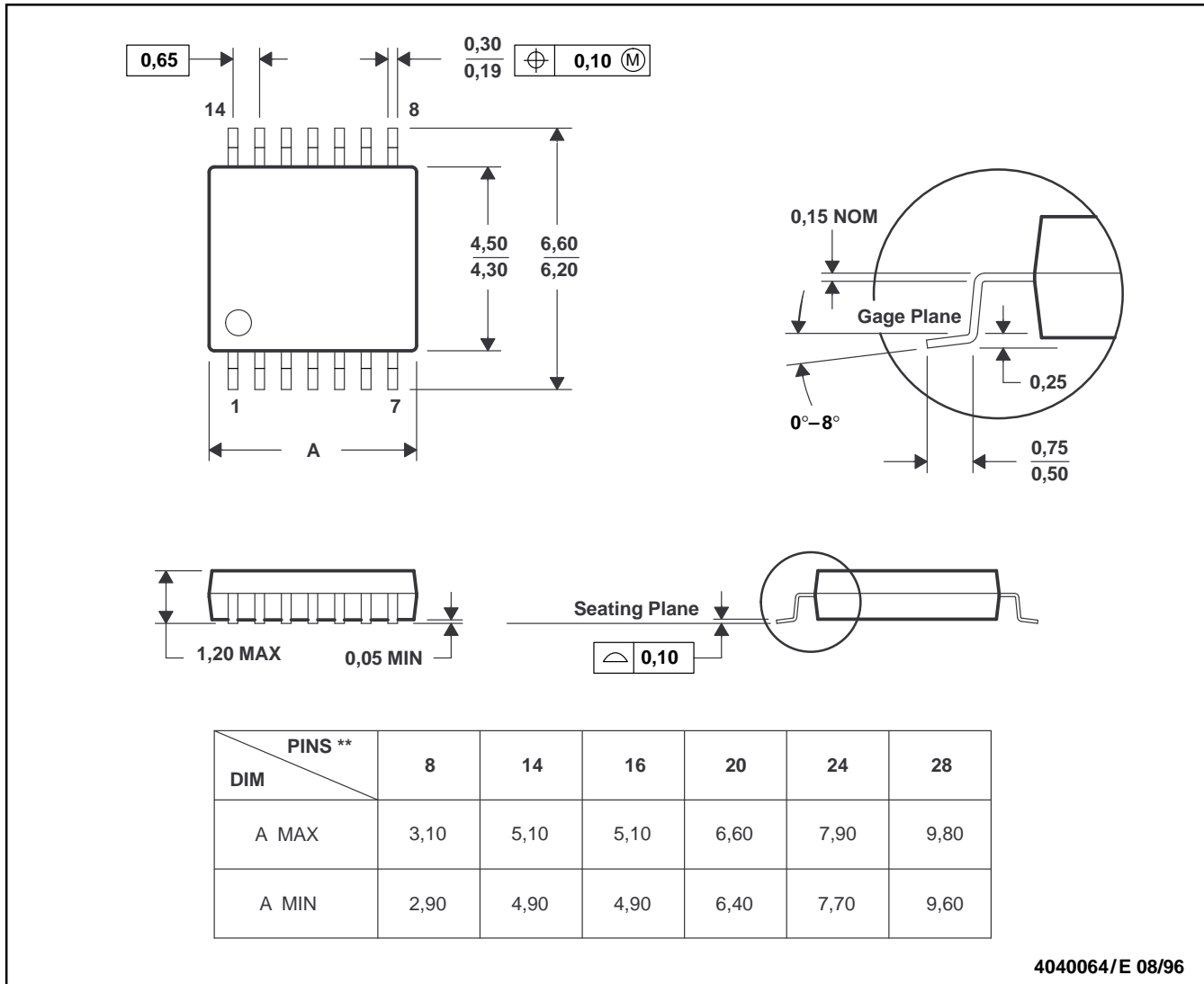
TLV5613
2.7 V TO 5.5 V 12-BIT PARALLEL DIGITAL-TO-ANALOG CONVERTER
WITH POWER DOWN

SLAS174A – DECEMBER 1997 – REVISED JUNE 1998

PW (R-PDSO-G)**

PLASTIC SMALL-OUTLINE PACKAGE

14 PIN SHOWN



4040064/E 08/96

- NOTES: A. All linear dimensions are in millimeters.
 B. This drawing is subject to change without notice.
 C. Body dimensions do not include mold flash or protrusion not to exceed 0,15.
 D. Falls within JEDEC MO-153